### 6.003: Signals and Systems

Z Transform

## Concept Map: Discrete-Time Systems

Multiple representations of DT systems.

## Block Diagram



System Functional

$$
\frac{Y}{X}=\mathcal{H}(\mathcal{R})=\frac{1}{1-\mathcal{R}-\mathcal{R}^{2}}
$$

Unit-Sample Response

$$
h[n]: \quad 1,1,2,3,5,8,13,21,34,55, \ldots
$$

Difference Equation
$y[n]=x[n]+y[n-1]+y[n-2]$

System Function

$$
H(z)=\frac{Y(z)}{X(z)}=\frac{z^{2}}{z^{2}-z-1}
$$

## Concept Map: Discrete-Time Systems

Relations among representations.

Block Diagram


Unit-Sample Response

$$
h[n]: \quad 1,1,2,3,5,8,13,21,34,55, \ldots
$$



Difference Equation

$$
y[n]=x[n]+y[n-1]+y[n-2]
$$

System Functional

$$
\frac{Y}{X}=\mathcal{H}(\mathcal{R})=\frac{1}{1-\mathcal{R}-\mathcal{R}^{2}}
$$



System Function

$$
H(z)=\frac{Y(z)}{X(z)}=\frac{z^{2}}{z^{2}-z-1}
$$

## Concept Map: Discrete-Time Systems

Two interpretations of "Delay."
Delay $\rightarrow \mathcal{R}$

Block Diagram


## System Functional

$$
\frac{Y}{X}=\mathcal{H}(\mathcal{R})=\frac{1}{1-\mathcal{R}-\mathcal{R}^{2}}
$$

Unit-Sample Response
index $\int_{\text {shift }} \quad h[n]: 1,1,2,3,5,8,13,21,34,55, \ldots$

Difference Equation

$$
y[n]=x[n]+y[n-1]+y[n-2]
$$

System Function

$$
H(z)=\frac{Y(z)}{X(z)}=\frac{z^{2}}{z^{2}-z-1}
$$

## Concept Map: Discrete-Time Systems

Relation between System Functional and System Function.

## Block Diagram



System Functional

$$
\frac{Y}{X}=\mathcal{H}(\mathcal{R})=\frac{1}{1-\mathcal{R}-\mathcal{R}^{2}}
$$



Unit-Sample Response

$$
h[n]: \quad 1,1,2,3,5,8,13,21,34,55, \ldots
$$

$$
\mathcal{R} \rightarrow \frac{1}{z}
$$

Difference Equation

$$
y[n]=x[n]+y[n-1]+y[n-2]
$$

System Function

$$
H(z)=\frac{Y(z)}{X(z)}=\frac{z^{2}}{z^{2}-z-1}
$$

## Check Yourself

What is relation of System Functional to Unit-Sample Response

## System Functional

$$
\frac{Y}{X}=\mathcal{H}(\mathcal{R})=\frac{1}{1-\mathcal{R}-\mathcal{R}^{2}}
$$



## Unit-Sample Response

$$
h[n]: \quad 1,1,2,3,5,8,13,21,34,55, \ldots
$$

Difference Equation

$$
y[n]=x[n]+y[n-1]+y[n-2]
$$

System Function

$$
H(z)=\frac{Y(z)}{X(z)}=\frac{z^{2}}{z^{2}-z-1}
$$

## Check Yourself

Expand functional in a series:

$$
\frac{Y}{X}=\mathcal{H}(\mathcal{R})=\frac{1}{1-\mathcal{R}-\mathcal{R}^{2}}
$$

$$
\begin{gathered}
1-\mathcal{R}-\mathcal{R}^{2} \begin{array}{c}
1+\mathcal{R}+2 \mathcal{R}^{2}+3 \mathcal{R}^{3}+5 \mathcal{R}^{4}+8 \mathcal{R}^{5} \\
\frac{1-\mathcal{R}-\mathcal{R}^{2}}{\mathcal{R}+\mathcal{R}^{2}} \\
\frac{\mathcal{R}-\mathcal{R}^{2}-\mathcal{R}^{3}}{2 \mathcal{R}^{2}+\mathcal{R}^{3}} \\
\frac{2 \mathcal{R}^{2}-2 \mathcal{R}^{3}-2 \mathcal{R}^{4}}{3 \mathcal{R}^{3}+2 \mathcal{R}^{4}} \\
\frac{3 \mathcal{R}^{3}-3 \mathcal{R}^{4}-3 \mathcal{R}^{5}}{\cdots}
\end{array} \\
\mathcal{H}(\mathcal{R})=\frac{1}{1-\mathcal{R}-\mathcal{R}^{2}}=1+\mathcal{R}+2 \mathcal{R}^{2}+3 \mathcal{R}^{3}+5 \mathcal{R}^{4}+8 \mathcal{R}^{5}+13 \mathcal{R}^{6}+\cdots
\end{gathered}
$$

## Check Yourself

Coefficients of series representation of $\mathcal{H}(\mathcal{R})$

$$
\mathcal{H}(\mathcal{R})=\frac{1}{1-\mathcal{R}-\mathcal{R}^{2}}=1+\mathcal{R}+2 \mathcal{R}^{2}+3 \mathcal{R}^{3}+5 \mathcal{R}^{4}+8 \mathcal{R}^{5}+13 \mathcal{R}^{6}+\cdots
$$

are the successive samples in the unit-sample response!

$$
h[n]: 1,1,2,3,5,8,13,21,34,55, \ldots
$$

If a system is composed of (only) adders, delays, and gains, then

$$
\begin{aligned}
\mathcal{H}(\mathcal{R}) & =h[0]+h[1] \mathcal{R}+h[2] \mathcal{R}^{2}+h[3] \mathcal{R}^{3}+h[4] \mathcal{R}^{4}+\cdots \\
& =\sum_{n} h[n] \mathcal{R}^{n}
\end{aligned}
$$

We can write the system function in terms of unit-sample response!

## Check Yourself

What is relation of System Functional to Unit-Sample Response?

## Block Diagram



$$
\frac{Y}{X}=\mathcal{H}(\mathcal{R})=\frac{1}{1-\mathcal{R}-\mathcal{R}^{2}}
$$

$$
\mathcal{H}(\mathcal{R})=\sum h[n] \mathcal{R}^{n}
$$

Unit-Sample Response

$$
h[n]: \quad 1,1,2,3,5,8,13,21,34,55, \ldots
$$

Difference Equation

$$
y[n]=x[n]+y[n-1]+y[n-2]
$$

System Function

$$
H(z)=\frac{Y(z)}{X(z)}=\frac{z^{2}}{z^{2}-z-1}
$$

## Check Yourself

What is relation of System Function to Unit-Sample Response?


## System Functional

$$
\frac{Y}{X}=\mathcal{H}(\mathcal{R})=\frac{1}{1-\mathcal{R}-\mathcal{R}^{2}}
$$

$$
\boldsymbol{\mathcal { H }}(\mathcal{R})=\sum h[n] \mathcal{R}^{n}
$$

Unit-Sample Response

$$
h[n]: \quad 1,1,2,3,5,8,13,21,34,55, \ldots
$$

Difference Equation

$$
y[n]=x[n]+y[n-1]+y[n-2]
$$

System Function

$$
H(z)=\frac{Y(z)}{X(z)}=\frac{z^{2}}{z^{2}-z-1}
$$

## Check Yourself

Start with the series expansion of system functional:

$$
\mathcal{H}(\mathcal{R})=\sum_{n} h[n] \mathcal{R}^{n}
$$

Substitute $\mathcal{R} \rightarrow \frac{1}{z}$ :

$$
H(z)=\sum_{n} h[n] z^{-n}
$$

## Check Yourself

What is relation of System Function to Unit-Sample Response?

## Block Diagram



Difference Equation

$$
y[n]=x[n]+y[n-1]+y[n-2]
$$

$$
\frac{Y}{X}=\mathcal{H}(\mathcal{R})=\frac{1}{1-\mathcal{R}-\mathcal{R}^{2}}
$$

Unit-Sample Response
System Functional

$$
\mathcal{H}(\mathcal{R})=\sum h[n] \mathcal{R}^{n}
$$

$$
h[n]: 1,1,2,3,5,8,13,21,34,55, \ldots
$$

$H(z)=\sum h[n] z^{-n}$

System Function

$$
H(z)=\frac{Y(z)}{X(z)}=\frac{z^{2}}{z^{2}-z-1}
$$

## Check Yourself

Start with the series expansion of system functional:

$$
\mathcal{H}(\mathcal{R})=\sum_{n} h[n] \mathcal{R}^{n}
$$

Substitute $\mathcal{R} \rightarrow \frac{1}{z}$ :

$$
H(z)=\sum_{n} h[n] z^{-n}
$$

Today: thinking about a system as a mathematical function $H(z)$ rather than as an operator.

## Z Transform

We call the relation between $H(z)$ and $h[n]$ the $Z$ transform.

$$
H(z)=\sum_{n} h[n] z^{-n}
$$

$Z$ transform maps a function of discrete time $n$ to a function of $z$.

Although motivated by system functions, we can define a $Z$ transform for any signal.

$$
X(z)=\sum_{n=-\infty}^{\infty} x[n] z^{-n}
$$

Notice that we include $n<0$ as well as $n>0 \rightarrow$ bilateral $Z$ transform (there is also a unilateral $Z$ transform with similar but not identical properties).

## Simple Z transforms

Find the $Z$ transform of the unit-sample signal.


$$
\begin{aligned}
& x[n]=\delta[n] \\
& X(z)=\sum_{n=-\infty}^{\infty} x[n] z^{-n}=x[0] z^{0}=1
\end{aligned}
$$

## Simple Z transforms

Find the $Z$ transform of a delayed unit-sample signal.


$$
\begin{aligned}
& x[n]=\delta[n-1] \\
& X(z)=\sum_{n=-\infty}^{\infty} x[n] z^{-n}=x[1] z^{-1}=z^{-1}
\end{aligned}
$$

## Check Yourself

## What is the $Z$ transform of the following signal.


$\begin{array}{lllll}\text { 1. } \frac{1}{1-\frac{7}{8} z} & \text { 2. } \frac{1}{1-\frac{7}{8} z^{-1}} & \text { 3. } \frac{z}{1-\frac{7}{8} z} & \text { 4. } \frac{z^{-1}}{1-\frac{7}{8} z^{-1}} & \text { 5. none }\end{array}$

## Check Yourself

What is the $Z$ transform of the following signal.

$$
\begin{aligned}
& x[n]=\left(\frac{7}{8}\right)^{n} u[n]
\end{aligned}
$$

$$
\begin{aligned}
& X(z)=\sum_{n=-\infty}^{\infty}\left(\frac{7}{8}\right)^{n} z^{-n} u[n]=\sum_{n=0}^{\infty}\left(\frac{7}{8}\right)^{n} z^{-n}=\frac{1}{1-\frac{7}{8} z^{-1}}
\end{aligned}
$$

## Check Yourself

## What is the $Z$ transform of the following signal. 2


$\begin{array}{lllll}\text { 1. } \frac{1}{1-\frac{7}{8} z} & \text { 2. } \frac{1}{1-\frac{7}{8} z^{-1}} & \text { 3. } \frac{z}{1-\frac{7}{8} z} & \text { 4. } \frac{z^{-1}}{1-\frac{7}{8} z^{-1}} & \text { 5. none }\end{array}$

## Z Transform Pairs

The signal $x[n]$, which is a function of time $n$, maps to a $\mathbf{Z}$ transform $X(z)$, which is a function of $z$.

$$
x[n]=\left(\frac{7}{8}\right)^{n} u[n] \quad \leftrightarrow \quad X(z)=\frac{1}{1-\frac{7}{8} z^{-1}}
$$

For what values of $z$ does $X(z)$ make sense?

The $Z$ transform is only defined for values of $z$ for which the defining sum converges.

$$
X(z)=\sum_{n=-\infty}^{\infty}\left(\frac{7}{8}\right)^{n} z^{-n} u[n]=\sum_{n=0}^{\infty}\left(\frac{7}{8}\right)^{n} z^{-n}=\frac{1}{1-\frac{7}{8} z^{-1}}
$$

Therefore $\left|\frac{7}{8} z^{-1}\right|<1$, i.e., $|z|>\frac{7}{8}$.

## Regions of Convergence

The Z transform $X(z)$ is a function of $z$ defined for all $z$ inside a Region of Convergence (ROC).

$$
x[n]=\left(\frac{7}{8}\right)^{n} u[n] \quad \leftrightarrow \quad X(z)=\frac{1}{1-\frac{7}{8} z^{-1}} ; \quad|z|>\frac{7}{8}
$$

$\mathrm{ROC}:|z|>\frac{7}{8}$

## Z Transform Mathematics

Based on properties of the $Z$ transform.

## Linearity:

| if | $x_{1}[n]$ | $\leftrightarrow$ | $X_{1}(z)$ | for $z$ in $\mathrm{ROC}_{1}$ |
| :--- | :---: | :---: | :---: | :--- |
| and | $x_{2}[n]$ | $\leftrightarrow$ | $X_{2}(z)$ | for $z$ in $\mathrm{ROC}_{2}$ |
| then | $x_{1}[n]+x_{2}[n]$ | $\leftrightarrow$ | $X_{1}(z)+X_{2}(z)$ | for $z$ in $\left(\mathrm{ROC}_{1} \cap \mathrm{ROC}_{2}\right)$. |

Let $y[n]=x_{1}[n]+x_{2}[n]$ then

$$
\begin{aligned}
Y(z) & =\sum_{n=-\infty}^{\infty} y[n] z^{-n} \\
& =\sum_{n=-\infty}^{\infty}\left(x_{1}[n]+x_{2}[n]\right) z^{-n} \\
& =\sum_{n=-\infty}^{\infty} x_{1}[n] z^{-n}+\sum_{n=-\infty}^{\infty} x_{2}[n] z^{-n} \\
& =X_{1}(z)+X_{2}(z)
\end{aligned}
$$

## Delay Property

If $x[n] \leftrightarrow X(z)$ for $z$ in ROC then $x[n-1] \leftrightarrow z^{-1} X(z)$ for $z$ in ROC.
We have already seen an example of this property.

$$
\begin{array}{rll}
\delta[n] & \leftrightarrow & 1 \\
\delta[n-1] & \leftrightarrow & z^{-1}
\end{array}
$$

More generally,

$$
X(z)=\sum_{n=-\infty}^{\infty} x[n] z^{-n}
$$

Let $y[n]=x[n-1]$ then

$$
Y(z)=\sum_{n=-\infty}^{\infty} y[n] z^{-n}=_{n=-\infty}^{\infty} x[n-1] z^{-n}
$$

Substitute $m=n-1$

$$
Y(z)=\sum_{m=-\infty}^{\infty} x[m] z^{-m-1}=z^{-1} X(z)
$$

## Rational Polynomials

A system that can be described by a linear difference equation with constant coefficients can also be described by a Z transform that is a ratio of polynomials in $z$.

$$
b_{0} y[n]+b_{1} y[n-1]+b_{2} y[n-2]+\cdots=a_{0} x[n]+a_{1} x[n-1]+a_{2} x[n-2]+\cdots
$$

Taking the $Z$ transform of both sides, and applying the delay property

$$
\begin{aligned}
& b_{0} Y(z)+b_{1} z^{-1} Y(z)+b_{2} z^{-2} Y(z)+\cdots=a_{0} X(z)+a_{1} z^{-1} X(z)+a_{2} z^{-2} X(z)+\cdots \\
& \begin{aligned}
H(z)=\frac{Y(z)}{X(z)} & =\frac{a_{0}+a_{1} z^{-1}+a_{2} z^{-2}+\cdots}{b_{0}+b_{1} z^{-1}+b_{2} z^{-2}+\cdots} \\
& =\frac{a_{0} z^{k}+a_{1} z^{k-1}+a_{2} z^{k-2}+\cdots}{b_{0} z^{k}+b_{1} z^{k-1}+b_{2} z^{k-2}+\cdots}
\end{aligned}
\end{aligned}
$$

## Rational Polynomials

Applying the fundamental theorem of algebra and the factor theorem, we can express the polynomials as a product of factors.

$$
\begin{aligned}
H(z) & =\frac{a_{0} z^{k}+a_{1} z^{k-1}+a_{2} z^{k-2}+\cdots}{b_{0} z^{k}+b_{1} z^{k-1}+b_{2} z^{k-2}+\cdots} \\
& =\frac{\left(z-z_{0}\right)\left(z-z_{1}\right) \cdots\left(z-z_{k}\right)}{\left(z-p_{0}\right)\left(z-p_{1}\right) \cdots\left(z-p_{k}\right)}
\end{aligned}
$$

where the roots are called poles and zeros.

## Rational Polynomials

Regions of convergence for $Z$ transform are delimited by circles in the Z-plane. The edges of the circles are at the poles.

Example: $x[n]=\alpha^{n} u[n]$

$$
\begin{aligned}
& \begin{aligned}
X(z) & =\sum_{n=-\infty}^{\infty} \alpha^{n} u[n] z^{-n}=\sum_{n=0}^{\infty} \alpha^{n} z^{-n} \\
& =\frac{1}{1-\alpha z^{-1}} ; \quad\left|\alpha z^{-1}\right|<1
\end{aligned} \\
& =\frac{z}{z-\alpha} ; \quad|z|>|\alpha|
\end{aligned}
$$



## Check Yourself

What DT signal has the following Z transform?

$$
\frac{z}{z-\frac{7}{8}} ;|z|<\frac{7}{8} \quad\left(\mathrm{RQC} \times \frac{\mathrm{C}}{\frac{7}{8}}\right.
$$

## Check Yourself

Recall that we already know a function whose $Z$ transform is the outer region.


What changes if the region changes?
The original sum

$$
X(z)=\sum_{n=0}^{\infty}\left(\frac{7}{8}\right)^{n} z^{-n}
$$

does not converge if $|z|<\frac{7}{8}$.

## Check Yourself

The functional form is still the same,

$$
H(z)=\frac{Y(z)}{X(z)}=\frac{z}{z-\frac{7}{8}}
$$

Therefore, the difference equation for this system is the same,

$$
y[n+1]-\frac{7}{8} y[n]=x[n+1] .
$$

Convergence inside $|z|=\frac{7}{8}$ corresponds to a left-sided (non-causal) response. Solve by iterating backwards in time:

$$
y[n]=\frac{8}{7}(y[n+1]-x[n+1])
$$

## Check Yourself

Solve by iterating backwards in time:

$$
y[n]=\frac{8}{7}(y[n+1]-x[n+1])
$$

Start "at rest":

| $n$ | $x[n]$ | $y[n]$ |
| :---: | :---: | :---: |
| $>0$ | 0 | 0 |
| 0 | 1 | 0 |
| -1 | 0 | $-\left(\frac{8}{7}\right)$ |
| -2 | 0 | $-\left(\frac{8}{7}\right)^{2}$ |
| -3 | 0 | $-\left(\frac{8}{7}\right)^{3}$ |
| $\ldots$ |  | $\cdots$ |
| $n$ |  | $-\left(\frac{8}{7}\right)^{-n}$ |

$$
y[n]=-\left(\frac{8}{7}\right)^{-n} ; \quad n<0=-\left(\frac{7}{8}\right)^{n} u[-1-n]
$$

## Check Yourself

Plot

$$
\begin{gathered}
y[n]=-\left(\frac{7}{8}\right)^{n} u[-1-n] \\
-1-3-2-1 \\
0
\end{gathered}
$$



## Check Yourself

What DT signal has the following Z transform?

$$
\frac{z}{z-\frac{7}{8}} ;|z|<\frac{7}{8} \quad \mathrm{RQC}^{-}
$$

$$
\begin{gathered}
y[n]=-\left(\frac{7}{8}\right)^{n} u[-1-n] \\
-4-\beta-2-1 \\
0
\end{gathered}
$$

## Check Yourself

Two signals and two regions of convergence.


## Check Yourself

Find the inverse transform of

$$
X(z)=\frac{-3 z}{2 z^{2}-5 z+2}
$$

given that the ROC includes the unit circle.

## Check Yourself

Find the inverse transform of

$$
X(z)=\frac{-3 z}{2 z^{2}-5 z+2}
$$

given that the ROC includes the unit circle.

Expand with partial fractions:

$$
X(z)=\frac{-3 z}{2 z^{2}-5 z+2}=\frac{1}{2 z-1}-\frac{2}{z-2}
$$

Not a standard form!

## Check Yourself

## Standard forms:



$$
y[n]=-\left(\frac{7}{8}\right)^{n} u[-1-n]
$$




## Check Yourself

Find the inverse transform of

$$
X(z)=\frac{-3 z}{2 z^{2}-5 z+2}
$$

given that the ROC includes the unit circle.

Expand with partial fractions:

$$
X(z)=\frac{-3 z}{2 z^{2}-5 z+2}=\frac{1}{2 z-1}-\frac{2}{z-2}
$$

Not a standard form!
Expand it differently: as a standard form:

$$
X(z)=\frac{-3 z}{2 z^{2}-5 z+2}=\frac{2 z}{2 z-1}-\frac{z}{z-2}=\frac{z}{z-\frac{1}{2}}-\frac{z}{z-2}
$$

Standard form: a pole at $\frac{1}{2}$ and a pole at 2 .

## Check Yourself

Ratio of polynomials in $z$ :

$$
X(z)=\frac{-3 z}{2 z^{2}-5 z+2}=\frac{z}{z-\frac{1}{2}}-\frac{z}{z-2}
$$

- a pole at $\frac{1}{2}$ and a pole at 2 .


Region of convergence is "outside" pole at $\frac{1}{2}$ but "inside" pole at 2 .

$$
x[n]=\left(\frac{1}{2}\right)^{n} u[n]+2^{n} u[-1-n]
$$

## Check Yourself

Plot.

$$
x[n]=\left(\frac{1}{2}\right)^{n} u[n]+2^{n} u[-1-n]
$$

## Check Yourself

Alternatively, stick with non-standard form:

$$
X(z)=\frac{-3 z}{2 z^{2}-5 z+2}=\frac{1}{2 z-1}-\frac{2}{z-2}
$$

Make it look more standard:

$$
X(z)=\frac{1}{2} z^{-1} \frac{z}{z-\frac{1}{2}}-2 z^{-1} \frac{z}{z-2}
$$

Now

$$
\begin{aligned}
& x[n]=\frac{1}{2} \mathcal{R}\left\{\left(\frac{1}{2}\right)^{n} u[n]\right\}+2 \mathcal{R}\left\{+2^{n} u[-1-n]\right\} \\
& =\frac{1}{2}\left\{\left(\frac{1}{2}\right)^{n-1} u[n-1]\right\}+2\left\{+2^{n-1} u[-n]\right\} \\
& =\left\{\left(\frac{1}{2}\right)^{n} u[n-1]\right\}+\left\{+2^{n} u[-n]\right\} \\
& x[n] \\
& 0 .
\end{aligned}
$$

## Check Yourself

Alternative 3: expand as polynomials in $z^{-1}$ :

$$
\begin{aligned}
X(z) & =\frac{-3 z}{2 z^{2}-5 z+2}=\frac{-3 z^{-1}}{2-5 z^{-1}+2 z^{-2}} \\
& =\frac{2}{2-z^{-1}}-\frac{1}{1-2 z^{-1}}=\frac{1}{1-\frac{1}{2} z^{-1}}-\frac{1}{1-2 z^{-1}}
\end{aligned}
$$

Now

$$
x[n]=\left(\frac{1}{2}\right)^{n} u[n]+2^{n} u[-1-n]
$$

## Check Yourself

Find the inverse transform of

$$
X(z)=\frac{-3 z}{2 z^{2}-5 z+2}
$$

given that the ROC includes the unit circle.


## Solving Difference Equations with Z Transforms

Start with difference equation:

$$
y[n]-\frac{1}{2} y[n-1]=\delta[n]
$$

Take the Z transform of this equation:

$$
Y(z)-\frac{1}{2} z^{-1} Y(z)=1
$$

Solve for $Y(z)$ :

$$
Y(z)=\frac{1}{1-\frac{1}{2} z^{-1}}
$$

Take the inverse $Z$ transform (by recognizing the form of the transform):

$$
y[n]=\left(\frac{1}{2}\right)^{n} u[n]
$$

## Inverse Z transform

The inverse $Z$ transform is defined by an integral that is not particularly easy to solve.

Formally,

$$
x[n]=\frac{1}{2 \pi j} \int_{C} X(z) z^{n-1} d z
$$

where $C$ represents a closed contour that circles the origin by running in a counterclockwise direction through the region of convergence. This integral is not generally easy to compute.

This equation can be useful to prove theorems.
There are better ways (e.g., partial fractions) to compute inverse transforms for the kinds of systems that we frequently encounter.

## Properties of Z Transforms

The use of $Z$ Transforms to solve differential equations depends on several important properties.
Property
$x[n]$
$X(z) \quad$ ROC
Linearity

$$
a X_{1}(z)+b X_{2}(z) \supset\left(R_{1} \cap R_{2}\right)
$$

Delay

$$
z^{-1} X(z)
$$

$$
R
$$

Multiply by $n$

$$
\begin{gathered}
a x_{1}[n]+b x_{2} \\
x[n-1]
\end{gathered}
$$

$$
n x[n]
$$

$$
-z \frac{d X(z)}{d z} \quad R
$$

Convolve in $n \sum_{m=-\infty}^{\infty} x_{1}[m] x_{2}[n-m] \quad X_{1}(z) X_{2}(z) \quad \supset\left(R_{1} \cap R_{2}\right)$

## Check Yourself

Find the inverse transform of $Y(z)=\left(\frac{z}{z-1}\right)^{2} ; \quad|z|>1$.

## Check Yourself

Find the inverse transform of $Y(z)=\left(\frac{z}{z-1}\right)^{2} ; \quad|z|>1$.
$y[n]$ corresponds to unit-sample response of the right-sided system

$$
\begin{aligned}
& \frac{Y}{X}=\left(\frac{z}{z-1}\right)^{2}=\left(\frac{1}{1-z^{-1}}\right)^{2}=\left(\frac{1}{1-\mathcal{R}}\right)^{2} \\
& =\left(1+\mathcal{R}+\mathcal{R}^{2}+\mathcal{R}^{3}+\cdots\right) \times\left(1+\mathcal{R}+\mathcal{R}^{2}+\mathcal{R}^{3}+\cdots\right) \\
& \begin{array}{clllll} 
& 1 & \mathcal{R} & \mathcal{R}^{2} & \mathcal{R}^{3} & \ldots \\
\hline 1 & 1 & \mathcal{R} & \mathcal{R}^{2} & \mathcal{R}^{3} & \ldots \\
\mathcal{R} & \mathcal{R} & \mathcal{R}^{2} & \mathcal{R}^{3} & \mathcal{R}^{4} & \ldots
\end{array} \\
& \begin{array}{llllll}
\mathcal{R}^{2} & \mathcal{R}^{2} & \mathcal{R}^{3} & \mathcal{R}^{4} & \mathcal{R}^{5} & \ldots
\end{array} \\
& \begin{array}{llllll}
\mathcal{R}^{3} & \mathcal{R}^{3} & \mathcal{R}^{4} & \mathcal{R}^{5} & \mathcal{R}^{6} & \ldots
\end{array} \\
& \frac{Y}{X}=1+2 \mathcal{R}+3 \mathcal{R}^{2}+4 \mathcal{R}^{3}+\cdots=\sum_{n=0}^{\infty}(n+1) \mathcal{R}^{n} \\
& y[n]=h[n]=(n+1) u[n]
\end{aligned}
$$

## Check Yourself

Table lookup method.

$$
\begin{aligned}
Y(z)=\left(\frac{z}{z-1}\right)^{2} & \leftrightarrow & y[n]=? \\
\frac{z}{z-1} & \leftrightarrow & u[n]
\end{aligned}
$$

## Properties of Z Transforms

The use of $Z$ Transforms to solve differential equations depends on several important properties.
Property
$x[n]$
$X(z) \quad$ ROC
Linearity

$$
a X_{1}(z)+b X_{2}(z) \supset\left(R_{1} \cap R_{2}\right)
$$

Delay

$$
z^{-1} X(z)
$$

$$
R
$$

Multiply by $n$

$$
\begin{gathered}
a x_{1}[n]+b x_{2} \\
x[n-1]
\end{gathered}
$$

$$
n x[n]
$$

$$
-z \frac{d X(z)}{d z} \quad R
$$

Convolve in $n \sum_{m=-\infty}^{\infty} x_{1}[m] x_{2}[n-m] \quad X_{1}(z) X_{2}(z) \quad \supset\left(R_{1} \cap R_{2}\right)$

## Check Yourself

Table lookup method.

$$
\begin{aligned}
& Y(z)=\left(\frac{z}{z-1}\right)^{2} \leftrightarrow y[n]=? \\
& \frac{z}{z-1} \leftrightarrow u[n] \\
& -z \frac{d}{d z}\left(\frac{z}{z-1}\right)=z\left(\frac{1}{z-1}\right)^{2} \leftrightarrow n u[n] \\
& z \times\left(-z \frac{d}{d z}\left(\frac{z}{z-1}\right)\right)=\left(\frac{z}{z-1}\right)^{2} \leftrightarrow \quad(n+1) u[n+1]=(n+1) u[n]
\end{aligned}
$$

## Concept Map: Discrete-Time Systems

Relations among representations.

Block Diagram


System Functional

$$
\frac{Y}{X}=\mathcal{H}(\mathcal{R})=\frac{1}{1-\mathcal{R}-\mathcal{R}^{2}}
$$



Unit-Sample Response
$h[n]: 1,1,2,3,5,8,13,21,34,55, \ldots$


System Function

$$
H(z)=\frac{Y(z)}{X(z)}=\frac{z^{2}}{z^{2}-z-1}
$$

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### 6.003 Signals and Systems

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