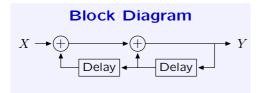
6.003: Signals and Systems

Z Transform

September 22, 2011

Multiple representations of DT systems.



System Functional

$$\frac{Y}{X} = \mathcal{H}(\mathcal{R}) = \frac{1}{1 - \mathcal{R} - \mathcal{R}^2}$$

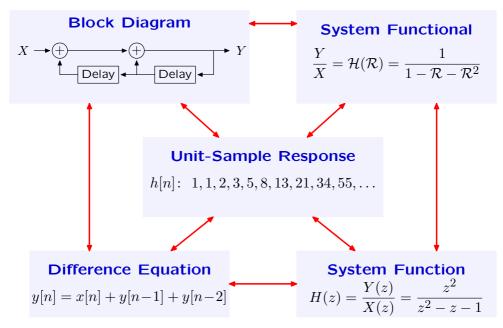
Unit-Sample Response

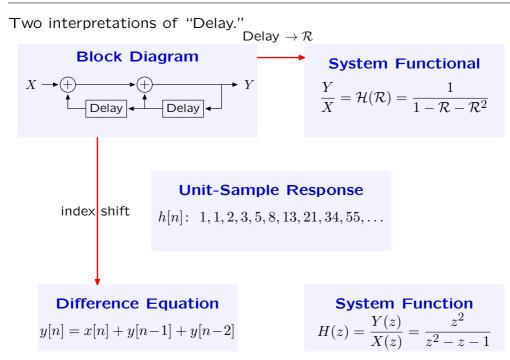
 $h[n]: 1, 1, 2, 3, 5, 8, 13, 21, 34, 55, \ldots$

Difference Equation y[n] = x[n] + y[n-1] + y[n-2]

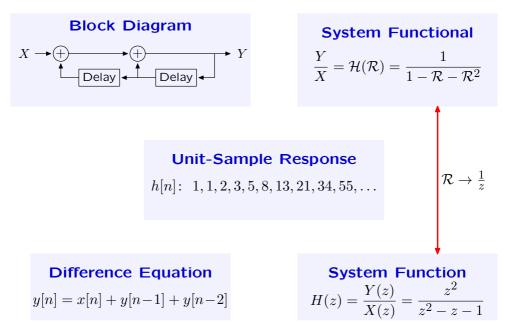
System Function
$$H(z) = \frac{Y(z)}{X(z)} = \frac{z^2}{z^2 - z - 1}$$

Relations among representations.

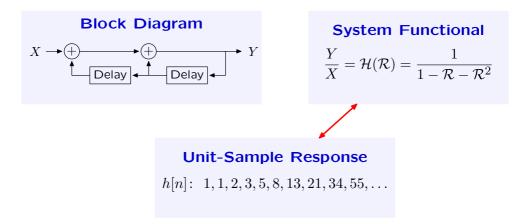




Relation between System Functional and System Function.



What is relation of System Functional to Unit-Sample Response



Difference Equation y[n] = x[n] + y[n-1] + y[n-2]

System Function $H(z) = \frac{Y(z)}{X(z)} = \frac{z^2}{z^2 - z - 1}$

Expand functional in a series:

$$\frac{Y}{X} = \mathcal{H}(\mathcal{R}) = \frac{1}{1 - \mathcal{R} - \mathcal{R}^2}$$

$$\mathcal{H}(\mathcal{R}) = \frac{1}{1 - \mathcal{R} - \mathcal{R}^2} = 1 + \mathcal{R} + 2\mathcal{R}^2 + 3\mathcal{R}^3 + 5\mathcal{R}^4 + 8\mathcal{R}^5 + 13\mathcal{R}^6 + \cdots$$

Coefficients of series representation of $\mathcal{H}(\mathcal{R})$

$$\mathcal{H}(\mathcal{R}) = \frac{1}{1 - \mathcal{R} - \mathcal{R}^2} = 1 + \mathcal{R} + 2\mathcal{R}^2 + 3\mathcal{R}^3 + 5\mathcal{R}^4 + 8\mathcal{R}^5 + 13\mathcal{R}^6 + \cdots$$

are the successive samples in the unit-sample response!

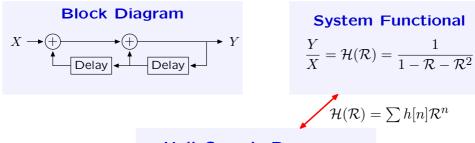
$$h[n]: 1, 1, 2, 3, 5, 8, 13, 21, 34, 55, \ldots$$

If a system is composed of (only) adders, delays, and gains, then

$$\mathcal{H}(\mathcal{R}) = h[0] + h[1]\mathcal{R} + h[2]\mathcal{R}^2 + h[3]\mathcal{R}^3 + h[4]\mathcal{R}^4 + \cdots$$
$$= \sum_n h[n]\mathcal{R}^n$$

We can write the system function in terms of unit-sample response!

What is relation of System Functional to Unit-Sample Response?



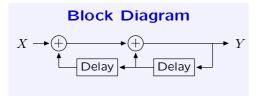
Unit-Sample Response

 $h[n]: 1, 1, 2, 3, 5, 8, 13, 21, 34, 55, \dots$

Difference Equation y[n] = x[n] + y[n-1] + y[n-2]

System Function
$$H(z) = \frac{Y(z)}{X(z)} = \frac{z^2}{z^2 - z - 1}$$

What is relation of System Function to Unit-Sample Response?



System Functional

$$\frac{Y}{X} = \mathcal{H}(\mathcal{R}) = \frac{1}{1 - \mathcal{R} - \mathcal{R}^2}$$

$$\mathcal{H}(\mathcal{R}) = \sum h[n]\mathcal{R}^n$$

Unit-Sample Response

 $h[n]: 1, 1, 2, 3, 5, 8, 13, 21, 34, 55, \ldots$

Difference Equation

 $y[n] = x[n] + y[n\!-\!1] + y[n\!-\!2]$

System Function $H(z) = \frac{Y(z)}{X(z)} = \frac{z^2}{z^2 - z - 1}$

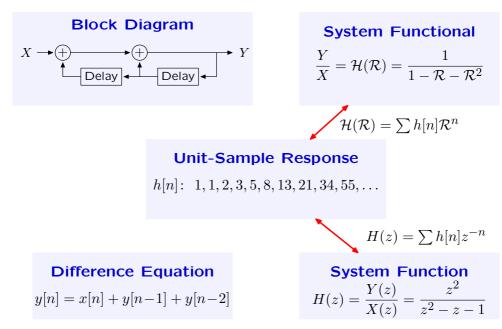
Start with the series expansion of system functional:

$$\mathcal{H}(\mathcal{R}) = \sum_n h[n]\mathcal{R}^n$$

Substitute $\mathcal{R} \to \frac{1}{z}$:

$$H(z) = \sum_{n} h[n] z^{-n}$$

What is relation of System Function to Unit-Sample Response?



Start with the series expansion of system functional:

$$\mathcal{H}(\mathcal{R}) = \sum_{n} h[n] \mathcal{R}^{n}$$

Substitute $\mathcal{R} \rightarrow \frac{1}{z}$:

$$H(z) = \sum_{n} h[n] z^{-n}$$

Today: thinking about a system as a **mathematical function** H(z) rather than as an operator.

Z Transform

We call the relation between H(z) and h[n] the Z transform.

$$H(z) = \sum_{n} h[n] z^{-n}$$

Z transform maps a function of discrete time n to a function of z.

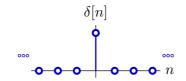
Although motivated by system functions, we can define a Z transform for any signal.

$$X(z) = \sum_{n = -\infty}^{\infty} x[n] z^{-n}$$

Notice that we include n < 0 as well as $n > 0 \rightarrow$ **bilateral** Z transform (there is also a unilateral Z transform with similar but not identical properties).

Simple Z transforms

Find the Z transform of the unit-sample signal.

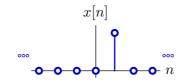


 $x[n] = \delta[n]$

$$X(z) = \sum_{n = -\infty}^{\infty} x[n]z^{-n} = x[0]z^{0} = 1$$

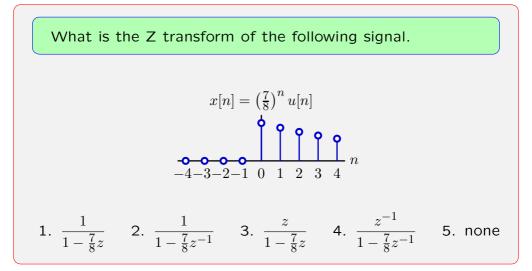
Simple Z transforms

Find the Z transform of a delayed unit-sample signal.



 $x[n] = \delta[n-1]$

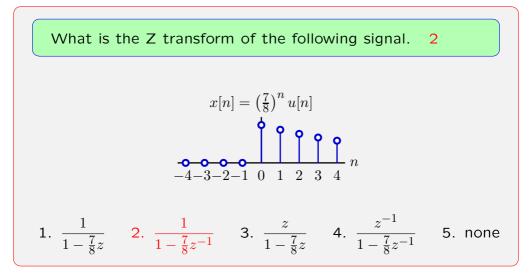
$$X(z) = \sum_{n=-\infty}^{\infty} x[n]z^{-n} = x[1]z^{-1} = z^{-1}$$



What is the Z transform of the following signal.

$$x[n] = \left(\frac{7}{8}\right)^n u[n]$$

$$X(z) = \sum_{n=-\infty}^{\infty} \left(\frac{7}{8}\right)^n z^{-n} u[n] = \sum_{n=0}^{\infty} \left(\frac{7}{8}\right)^n z^{-n} = \frac{1}{1 - \frac{7}{8}z^{-1}}$$



Z Transform Pairs

The signal x[n], which is a function of time n, maps to a Z transform X(z), which is a function of z.

$$x[n] = \left(\frac{7}{8}\right)^n u[n] \quad \leftrightarrow \quad X(z) = \frac{1}{1 - \frac{7}{8}z^{-1}}$$

For what values of z does X(z) make sense?

The Z transform is only defined for values of z for which the defining sum converges.

$$X(z) = \sum_{n=-\infty}^{\infty} \left(\frac{7}{8}\right)^n z^{-n} u[n] = \sum_{n=0}^{\infty} \left(\frac{7}{8}\right)^n z^{-n} = \frac{1}{1 - \frac{7}{8}z^{-1}}$$

herefore $\left|\frac{7}{8}z^{-1}\right| < 1$, i.e., $|z| > \frac{7}{8}$.

Regions of Convergence

The Z transform X(z) is a function of z defined for all z inside a **Region of Convergence (ROC)**.

$$x[n] = \left(\frac{7}{8}\right)^n u[n] \quad \leftrightarrow \quad X(z) = \frac{1}{1 - \frac{7}{8}z^{-1}}; \quad |z| > \frac{7}{8}$$

ROC: $|z| > \frac{7}{8}$

Z Transform Mathematics

Based on properties of the Z transform.

Linearity:

 $\begin{array}{lll} \text{if} & x_1[n] & \leftrightarrow & X_1(z) & \text{for } z \text{ in } \mathsf{ROC}_1 \\ \text{and} & x_2[n] & \leftrightarrow & X_2(z) & \text{for } z \text{ in } \mathsf{ROC}_2 \\ \text{then} & x_1[n] + x_2[n] & \leftrightarrow & X_1(z) + X_2(z) & \text{for } z \text{ in } (\mathsf{ROC}_1 \cap \mathsf{ROC}_2). \end{array}$

Let
$$y[n] = x_1[n] + x_2[n]$$
 then

$$Y(z) = \sum_{n=-\infty}^{\infty} y[n]z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} (x_1[n] + x_2[n])z^{-n}$$

$$= \sum_{n=-\infty}^{\infty} x_1[n]z^{-n} + \sum_{n=-\infty}^{\infty} x_2[n]z^{-n}$$

$$= X_1(z) + X_2(z)$$

Delay Property

If $x[n] \leftrightarrow X(z)$ for z in ROC then $x[n-1] \leftrightarrow z^{-1}X(z)$ for z in ROC.

We have already seen an example of this property.

$$\begin{array}{cccc} \delta[n] & \leftrightarrow & 1 \\ \delta[n-1] & \leftrightarrow & z^{-1} \end{array}$$

More generally,

$$X(z) = \sum_{n = -\infty}^{\infty} x[n] z^{-n}$$

Let
$$y[n] = x[n-1]$$
 then

$$Y(z) = \sum_{n=-\infty}^{\infty} y[n]z^{-n} = \sum_{n=-\infty}^{\infty} x[n-1]z^{-n}$$

Substitute m = n - 1

$$Y(z) = \sum_{m=-\infty}^{\infty} x[m] z^{-m-1} = z^{-1} X(z)$$

Rational Polynomials

A system that can be described by a linear difference equation with constant coefficients can also be described by a Z transform that is a **ratio of polynomials** in z.

$$b_0y[n] + b_1y[n-1] + b_2y[n-2] + \dots = a_0x[n] + a_1x[n-1] + a_2x[n-2] + \dots$$

Taking the Z transform of both sides, and applying the delay property

$$b_0 Y(z) + b_1 z^{-1} Y(z) + b_2 z^{-2} Y(z) + \dots = a_0 X(z) + a_1 z^{-1} X(z) + a_2 z^{-2} X(z) + \dots$$
$$H(z) = \frac{Y(z)}{X(z)} = \frac{a_0 + a_1 z^{-1} + a_2 z^{-2} + \dots}{b_0 + b_1 z^{-1} + b_2 z^{-2} + \dots}$$
$$= \frac{a_0 z^k + a_1 z^{k-1} + a_2 z^{k-2} + \dots}{b_0 z^k + b_1 z^{k-1} + b_2 z^{k-2} + \dots}$$

Rational Polynomials

Applying the fundamental theorem of algebra and the factor theorem, we can express the polynomials as a product of factors.

$$H(z) = \frac{a_0 z^k + a_1 z^{k-1} + a_2 z^{k-2} + \cdots}{b_0 z^k + b_1 z^{k-1} + b_2 z^{k-2} + \cdots}$$
$$= \frac{(z - z_0) \ (z - z_1) \ \cdots \ (z - z_k)}{(z - p_0) \ (z - p_1) \ \cdots \ (z - p_k)}$$

where the roots are called **poles** and **zeros**.

Rational Polynomials

Regions of convergence for Z transform are delimited by circles in the Z-plane. The edges of the circles are at the poles.

Example: $x[n] = \alpha^n u[n]$

$$X(z) = \sum_{n=-\infty}^{\infty} \alpha^{n} u[n] z^{-n} = \sum_{n=0}^{\infty} \alpha^{n} z^{-n}$$

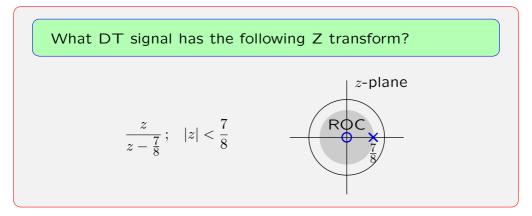
= $\frac{1}{1 - \alpha z^{-1}}; \quad \left| \alpha z^{-1} \right| < 1$
= $\frac{z}{z - \alpha}; \quad |z| > |\alpha|$
$$x[n] = \alpha^{n} u[n]$$

$$\underbrace{z}_{z - \alpha}$$

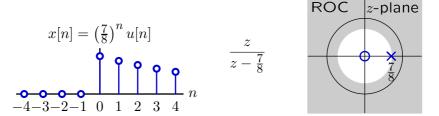
ROC

z-plane

 $\frac{\mathbf{x}}{\alpha}$



Recall that we already know a function whose Z transform is the outer region.



What changes if the region changes?

The original sum

$$X(z) = \sum_{n=0}^{\infty} \left(\frac{7}{8}\right)^n z^{-n}$$

does not converge if $|z| < \frac{7}{8}$.

The functional form is still the same,

$$H(z) = \frac{Y(z)}{X(z)} = \frac{z}{z - \frac{7}{8}}$$

Therefore, the difference equation for this system is the same, $y[n+1] - \frac{7}{8}y[n] = x[n+1]\,.$

Convergence **inside** $|z| = \frac{7}{8}$ corresponds to a left-sided (non-causal) response. Solve by iterating backwards in time:

$$y[n] = \frac{8}{7} (y[n+1] - x[n+1])$$

Solve by iterating backwards in time:

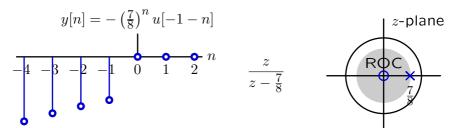
$$y[n] = \frac{8}{7} (y[n+1] - x[n+1])$$

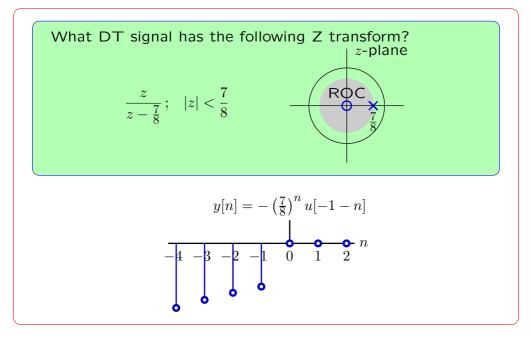
Start "at rest":

| n | x[n] | y[n] |
|-----|------|----------------------------------|
| > 0 | 0 | 0 |
| 0 | 1 | 0 |
| -1 | 0 | $-(\frac{8}{7})$ |
| -2 | 0 | $-\left(\frac{8}{7}\right)^{2}$ |
| -3 | 0 | $-\left(\frac{8}{7}\right)^3$ |
| ••• | | ••• |
| n | | $-\left(\frac{8}{7}\right)^{-n}$ |

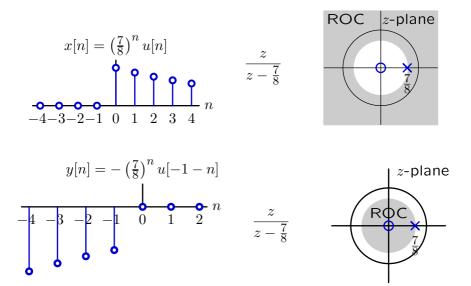
$$y[n] = -\left(\frac{8}{7}\right)^{-n}; \quad n < 0 = -\left(\frac{7}{8}\right)^{n} u[-1-n]$$

Plot





Two signals and two regions of convergence.



Find the inverse transform of
$$X(z) = \frac{-3z}{2z^2 - 5z + 2}$$
 given that the ROC includes the unit circle.

Find the inverse transform of $X(z) = \frac{-3z}{2z^2 - 5z + 2}$

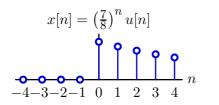
given that the ROC includes the unit circle.

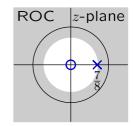
Expand with partial fractions:

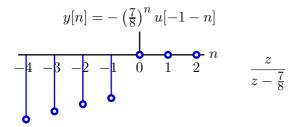
$$X(z) = \frac{-3z}{2z^2 - 5z + 2} = \frac{1}{2z - 1} - \frac{2}{z - 2}$$

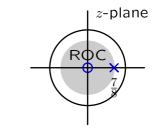
Not a standard form!

Standard forms:









 $\frac{z}{z-\frac{7}{8}}$

Find the inverse transform of $X(z) = \frac{-3z}{2z^2 - 5z + 2}$

given that the ROC includes the unit circle.

Expand with partial fractions:

$$X(z) = \frac{-3z}{2z^2 - 5z + 2} = \frac{1}{2z - 1} - \frac{2}{z - 2}$$

Not a standard form!

Expand it differently: as a standard form:

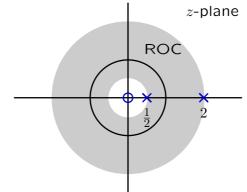
$$X(z) = \frac{-3z}{2z^2 - 5z + 2} = \frac{2z}{2z - 1} - \frac{z}{z - 2} = \frac{z}{z - \frac{1}{2}} - \frac{z}{z - 2}$$

Standard form: a pole at $\frac{1}{2}$ and a pole at 2.

Ratio of polynomials in z:

$$X(z) = \frac{-3z}{2z^2 - 5z + 2} = \frac{z}{z - \frac{1}{2}} - \frac{z}{z - 2}$$

- a pole at $\frac{1}{2}$ and a pole at 2.



Region of convergence is "outside" pole at $\frac{1}{2}$ but "inside" pole at 2. $x[n] = \left(\frac{1}{2}\right)^n u[n] + 2^n u[-1-n]$

Alternatively, stick with non-standard form:

$$X(z) = \frac{-3z}{2z^2 - 5z + 2} = \frac{1}{2z - 1} - \frac{2}{z - 2}$$

Make it look more standard:

$$X(z) = \frac{1}{2}z^{-1}\frac{z}{z-\frac{1}{2}} - 2z^{-1}\frac{z}{z-2}$$

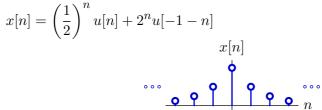
Now

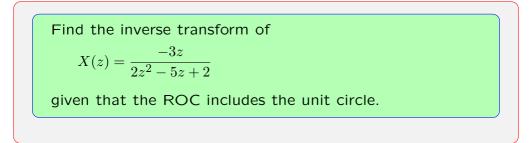
$$x[n] = \frac{1}{2} \mathcal{R} \left\{ \left(\frac{1}{2}\right)^n u[n] \right\} + 2 \mathcal{R} \left\{ +2^n u[-1-n] \right\}$$
$$= \frac{1}{2} \left\{ \left(\frac{1}{2}\right)^{n-1} u[n-1] \right\} + 2 \left\{ +2^{n-1} u[-n] \right\}$$
$$= \left\{ \left(\frac{1}{2}\right)^n u[n-1] \right\} + \left\{ +2^n u[-n] \right\}$$
$$x[n]$$

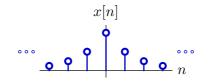
Alternative 3: expand as polynomials in z^{-1} :

$$X(z) = \frac{-3z}{2z^2 - 5z + 2} = \frac{-3z^{-1}}{2 - 5z^{-1} + 2z^{-2}}$$
$$= \frac{2}{2 - z^{-1}} - \frac{1}{1 - 2z^{-1}} = \frac{1}{1 - \frac{1}{2}z^{-1}} - \frac{1}{1 - 2z^{-1}}$$

Now







Solving Difference Equations with Z Transforms

Start with difference equation:

$$y[n] - \frac{1}{2}y[n-1] = \delta[n]$$

Take the Z transform of this equation:

$$Y(z) - \frac{1}{2}z^{-1}Y(z) = 1$$

Solve for Y(z):

$$Y(z) = \frac{1}{1 - \frac{1}{2}z^{-1}}$$

Take the inverse Z transform (by recognizing the form of the transform):

$$y[n] = \left(\frac{1}{2}\right)^n u[n]$$

Inverse Z transform

The inverse Z transform is defined by an integral that is not particularly easy to solve.

Formally,

$$x[n] = \frac{1}{2\pi j} \int_C X(z) z^{n-1} dz$$

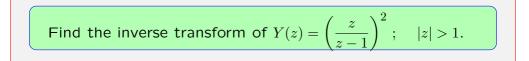
where C represents a closed contour that circles the origin by running in a counterclockwise direction through the region of convergence. This integral is not generally easy to compute.

This equation can be useful to prove theorems.

There are better ways (e.g., partial fractions) to compute inverse transforms for the kinds of systems that we frequently encounter.

Properties of Z Transforms

The use of Z Transforms to solve differential equations depends on several important properties.



Find the inverse transform of
$$Y(z) = \left(\frac{z}{z-1}\right)^2$$
; $|z| > 1$.

y[n] corresponds to unit-sample response of the **right-sided** system

$$\frac{Y}{X} = \left(\frac{z}{z-1}\right)^2 = \left(\frac{1}{1-z^{-1}}\right)^2 = \left(\frac{1}{1-\mathcal{R}}\right)^2$$
$$= \left(1+\mathcal{R}+\mathcal{R}^2+\mathcal{R}^3+\cdots\right) \times \left(1+\mathcal{R}+\mathcal{R}^2+\mathcal{R}^3+\cdots\right)$$
$$\frac{1}{1} \frac{\mathcal{R}}{\mathcal{R}} \frac{\mathcal{R}^2}{\mathcal{R}^3} \frac{\mathcal{R}^3}{\mathcal{R}^3} \cdots$$
$$\frac{\mathcal{R}}{\mathcal{R}} \mathcal{R} \frac{\mathcal{R}^2}{\mathcal{R}^3} \frac{\mathcal{R}^3}{\mathcal{R}^4} \frac{\mathcal{R}^5}{\mathcal{R}^5} \cdots$$
$$\frac{\mathcal{R}^3}{\mathcal{R}^3} \frac{\mathcal{R}^4}{\mathcal{R}^4} \frac{\mathcal{R}^5}{\mathcal{R}^5} \frac{\mathcal{R}^6}{\mathcal{R}^6} \cdots$$
$$\frac{Y}{X} = 1+2\mathcal{R}+3\mathcal{R}^2+4\mathcal{R}^3+\cdots = \sum_{n=0}^{\infty} (n+1)\mathcal{R}^n$$
$$y[n] = h[n] = (n+1)u[n]$$

Table lookup method.

$$Y(z) = \left(\frac{z}{z-1}\right)^2 \quad \leftrightarrow \quad y[n] = ?$$
$$\frac{z}{z-1} \quad \leftrightarrow \quad u[n]$$

Properties of Z Transforms

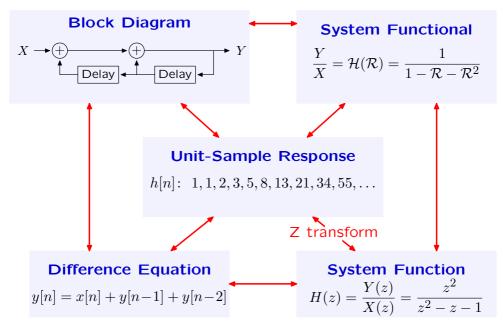
The use of Z Transforms to solve differential equations depends on several important properties.

Table lookup method.

$$Y(z) = \left(\frac{z}{z-1}\right)^2 \quad \leftrightarrow \quad y[n] = ?$$
$$\frac{z}{z-1} \quad \leftrightarrow \quad u[n]$$
$$-z\frac{d}{dz}\left(\frac{z}{z-1}\right) = z\left(\frac{1}{z-1}\right)^2 \quad \leftrightarrow \quad nu[n]$$
$$z \times \left(-z\frac{d}{dz}\left(\frac{z}{z-1}\right)\right) = \left(\frac{z}{z-1}\right)^2 \quad \leftrightarrow \quad (n+1)u[n+1] = (n+1)u[n]$$

Concept Map: Discrete-Time Systems

Relations among representations.



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